

Inductive Network Model for the Radiation Analysis of Electrically Small Parallel-Plate Structures

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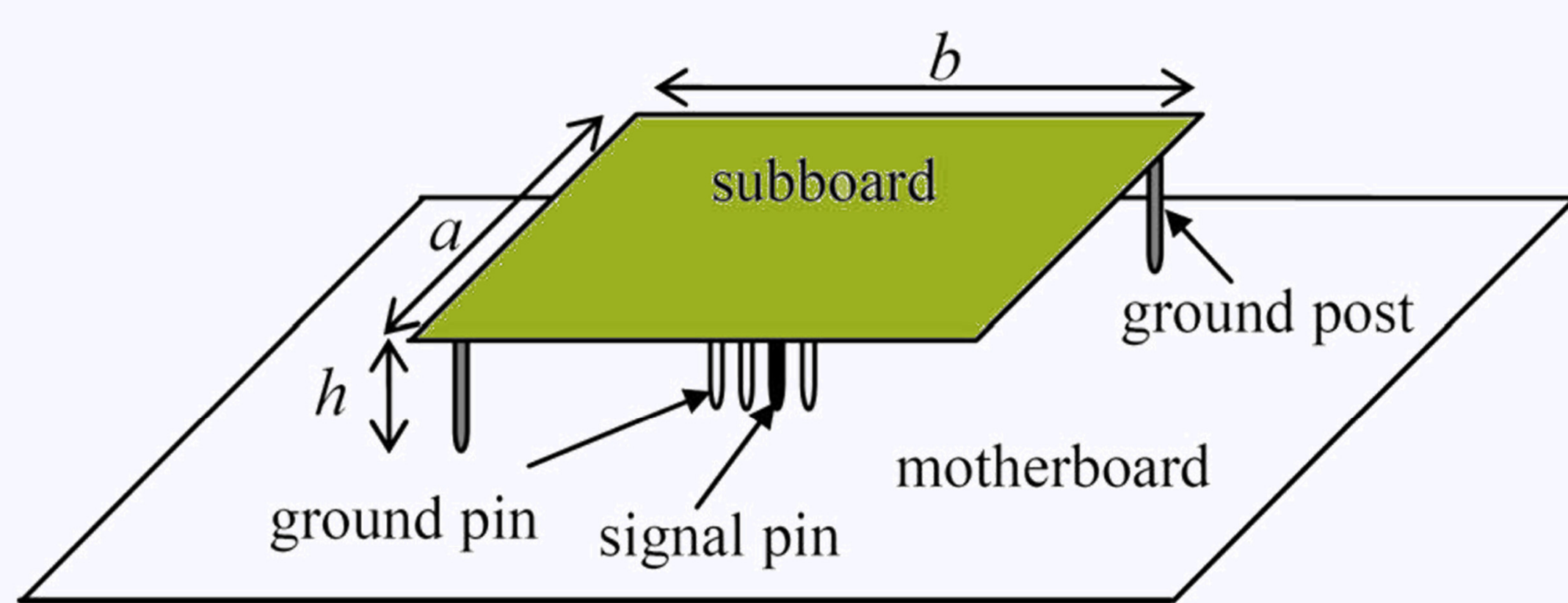
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INTRODUCTION

Vertical interconnections between electrically small structures as often encountered in motherboard subboard applications are source of radiated emissions

Modeling approach:

- equivalent circuit for interconnections
- hertzian-dipole radiation characteristic

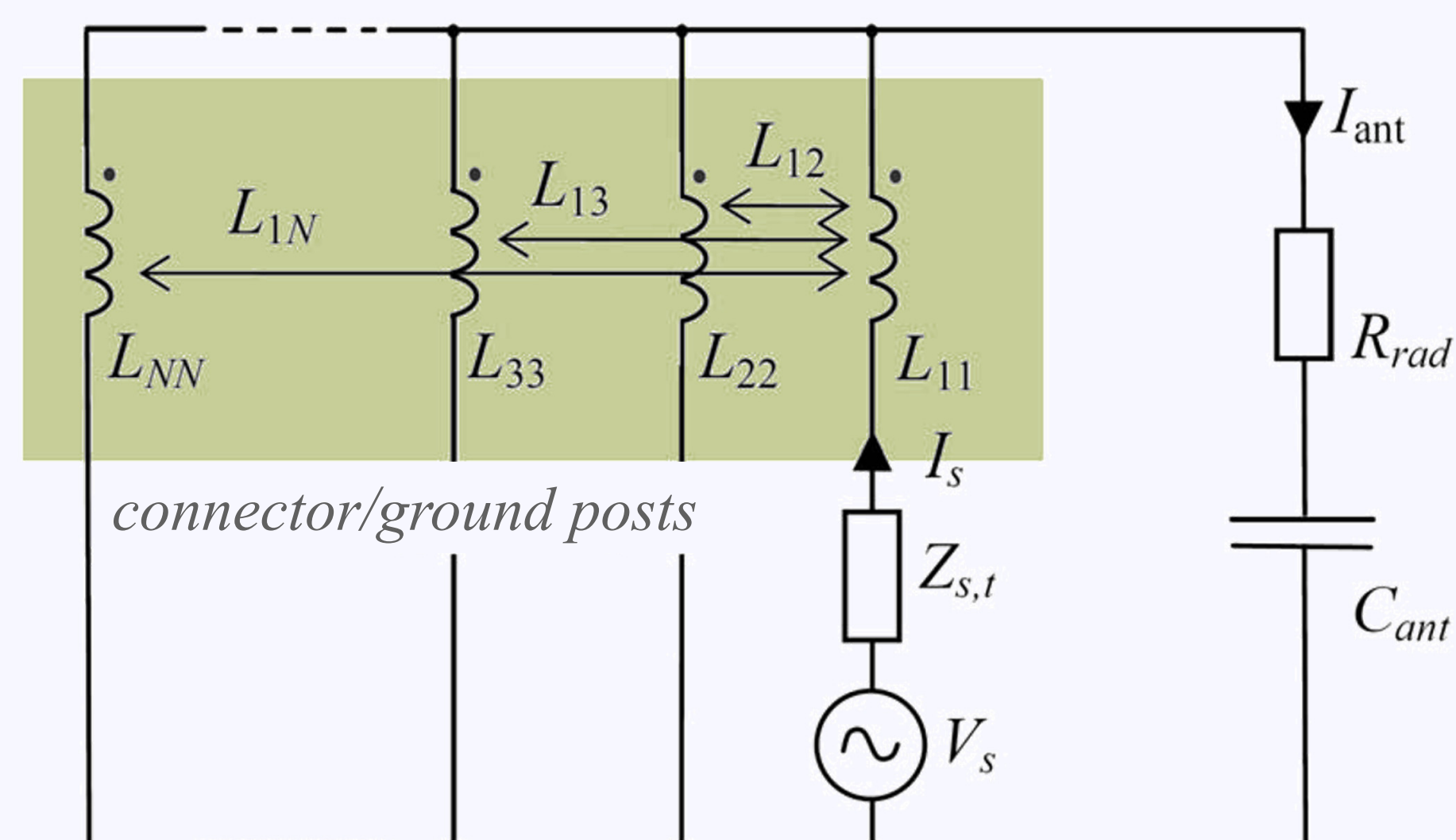


Electrical small parallel-plate structure ($\lambda \gg a, b \gg h$)

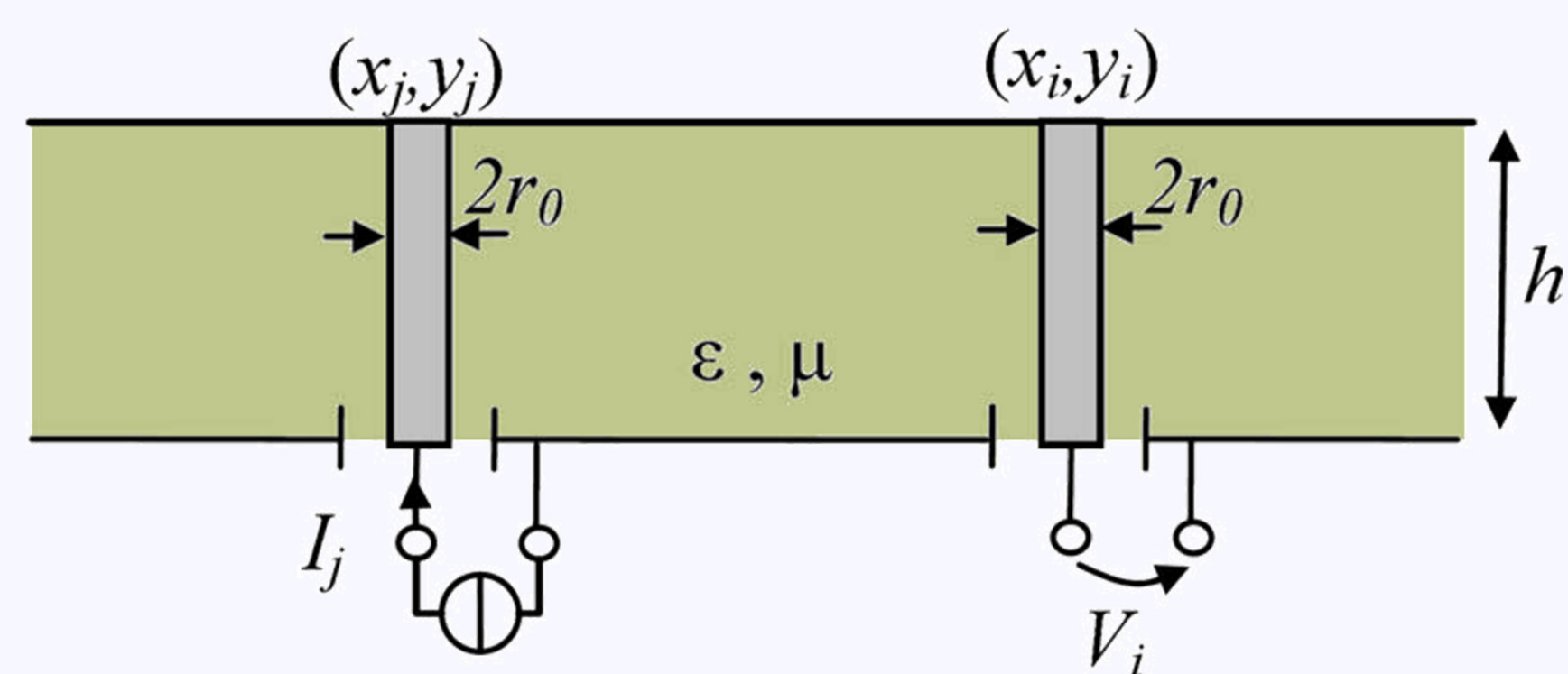
QUASISTATIC GREENS FUNCTION

$$G(r|r_s) = j\omega \frac{\mu h}{ab} \sum_{n=0}^{\infty} \sum_{m=0}^{\infty} \frac{K_{m,n}}{k_{xm}^2 + k_{yn}^2} \times \cos(k_{xm} x) \cos(k_{yn} y) \cos(k_{xm} x_s) \cos(k_{yn} y_s)$$

LOW FREQUENCY EQUIVALENT-CIRCUIT MODEL



NETWORK ELEMENTS



Side view of source and observation ports with circular geometry and radius r_0

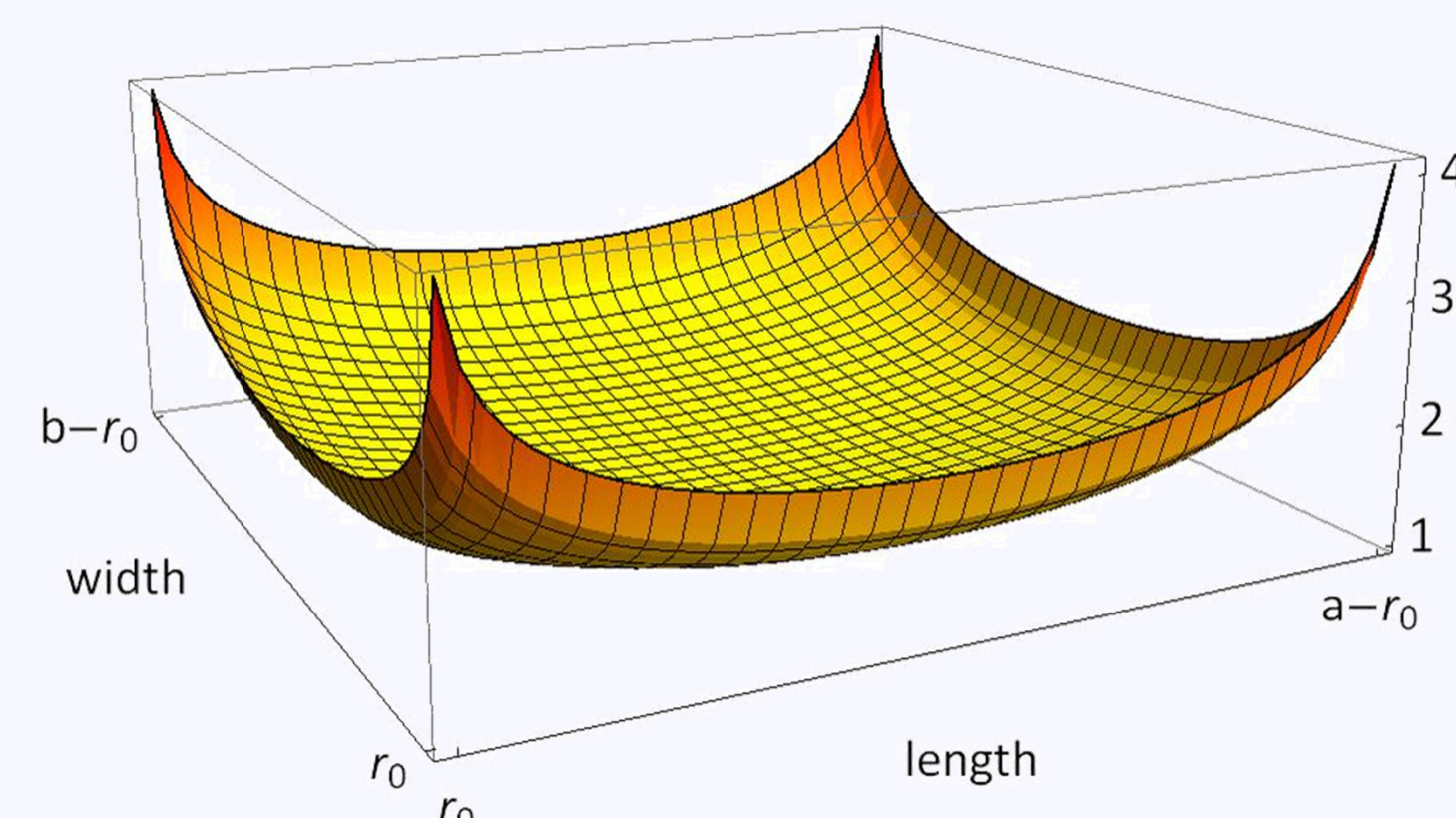
Self-inductance

$$L_{ii} = -\frac{h\mu}{4\pi} \left(\begin{aligned} &\ln(\cosh(2\pi y_i/a) - \cos(2\pi x_i/a)) \\ &+ \ln(\cosh(2\pi(b-y_i)/a) - \cos(2\pi x_i/a)) \\ &+ 2\ln(\sinh(\pi y_i/a) \sinh(\pi(b-y_i)/a)) \\ &+ 2\ln(\sin(\pi x_i/a)) + 2\ln(\pi r_0/a) \\ &+ 8\ln(2) - \frac{b}{a} \frac{13}{3} \pi - 4\pi \frac{(y_i - b/2)^2}{ba} \end{aligned} \right)$$

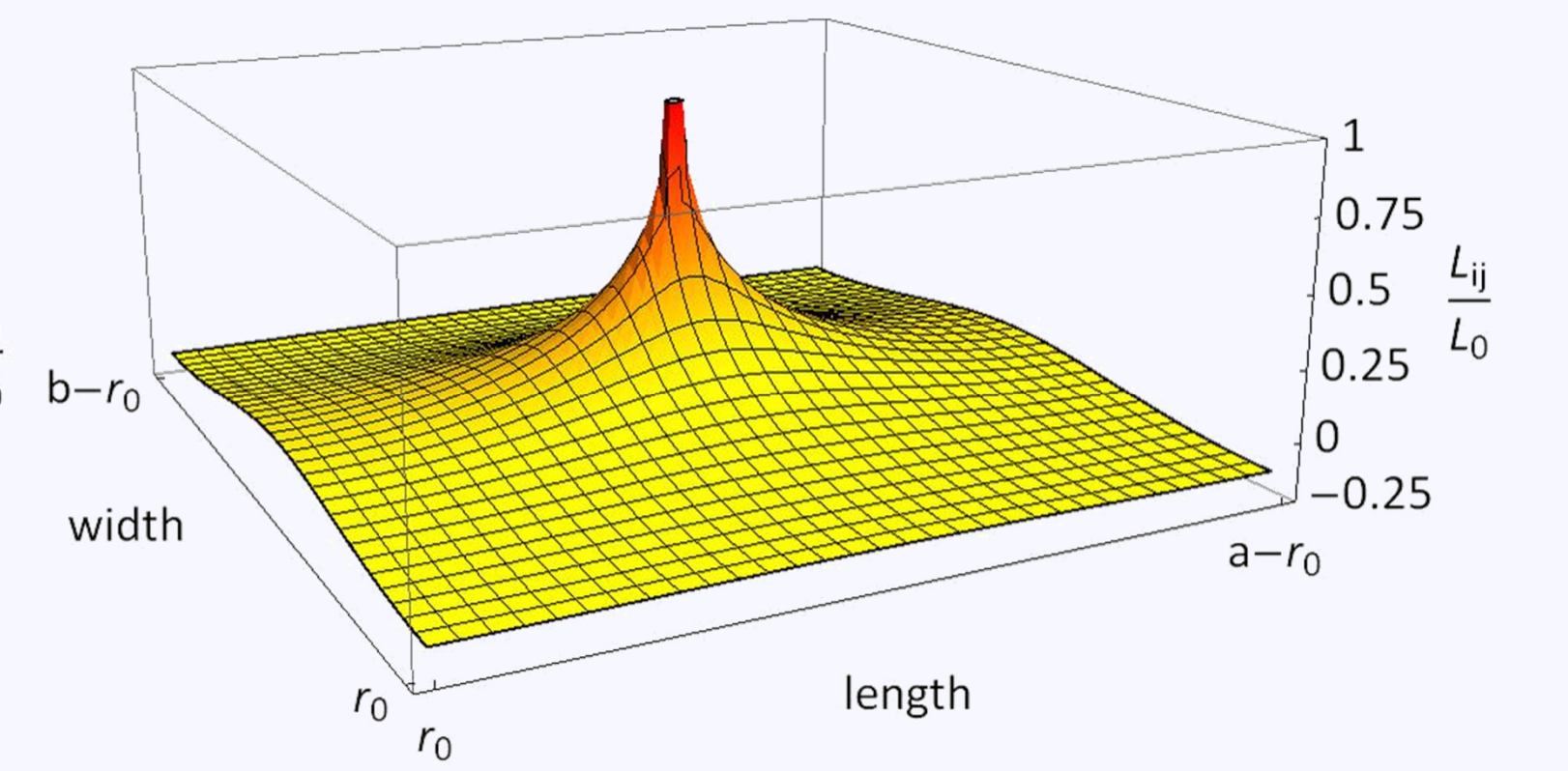
Mutual-inductance

$$L_{ij} = -\frac{\mu h}{4\pi} \left(\begin{aligned} &W(y_i - y_j, x_i, x_j) + W(2b - (y_i + y_j), x_i, x_j) \\ &+ W(2b - (y_i - y_j), x_i, x_j) + W(y_i + y_j, x_i, x_j) \end{aligned} \right) + \mu h \left(\frac{y_i^2 + y_j^2 - y_i}{2ab} - \frac{y_i}{a} \right) - \frac{\mu h}{\pi} \ln(4) + \mu h \frac{b}{a} \frac{7}{3}$$

$$\text{with } W(y', x_i, x_j) = \ln \left(\begin{aligned} &(\cosh(\pi y'/a) - \cos((x_i + x_j)\pi/a)) \\ &\times (\cosh(\pi y'/a) - \cos((x_i - x_j)\pi/a)) \end{aligned} \right)$$



Normalized self inductance, depending on the port's position



Normalized mutual inductance to a port in the plates center

Plate capacitance

$$C_{ant} = \epsilon \frac{ab}{h} \left(1 + \frac{2h}{\pi a} \left(1 + \ln\left(\frac{\pi a}{h}\right) \right) \right) \times \left(1 + \frac{2h}{\pi b} \left(1 + \ln\left(\frac{\pi b}{h}\right) \right) \right)$$

Radiation resistance

Hertzian dipole above infinite plane

$$R_{rad} = 40\pi^2 \left(\frac{2h}{c_0} f \right)^2$$

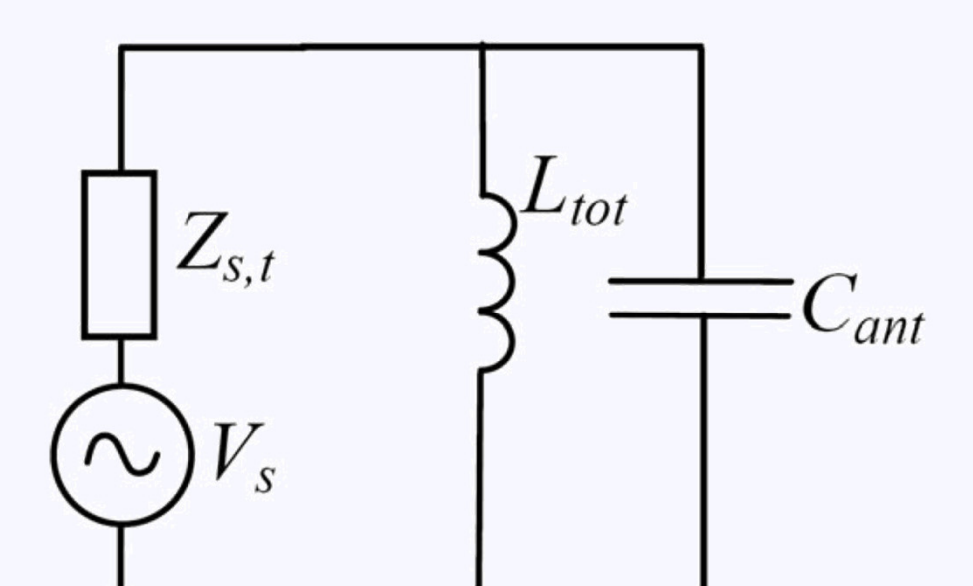
RESONANCE FREQUENCY ESTIMATION

Parallel resonance of signal return path:

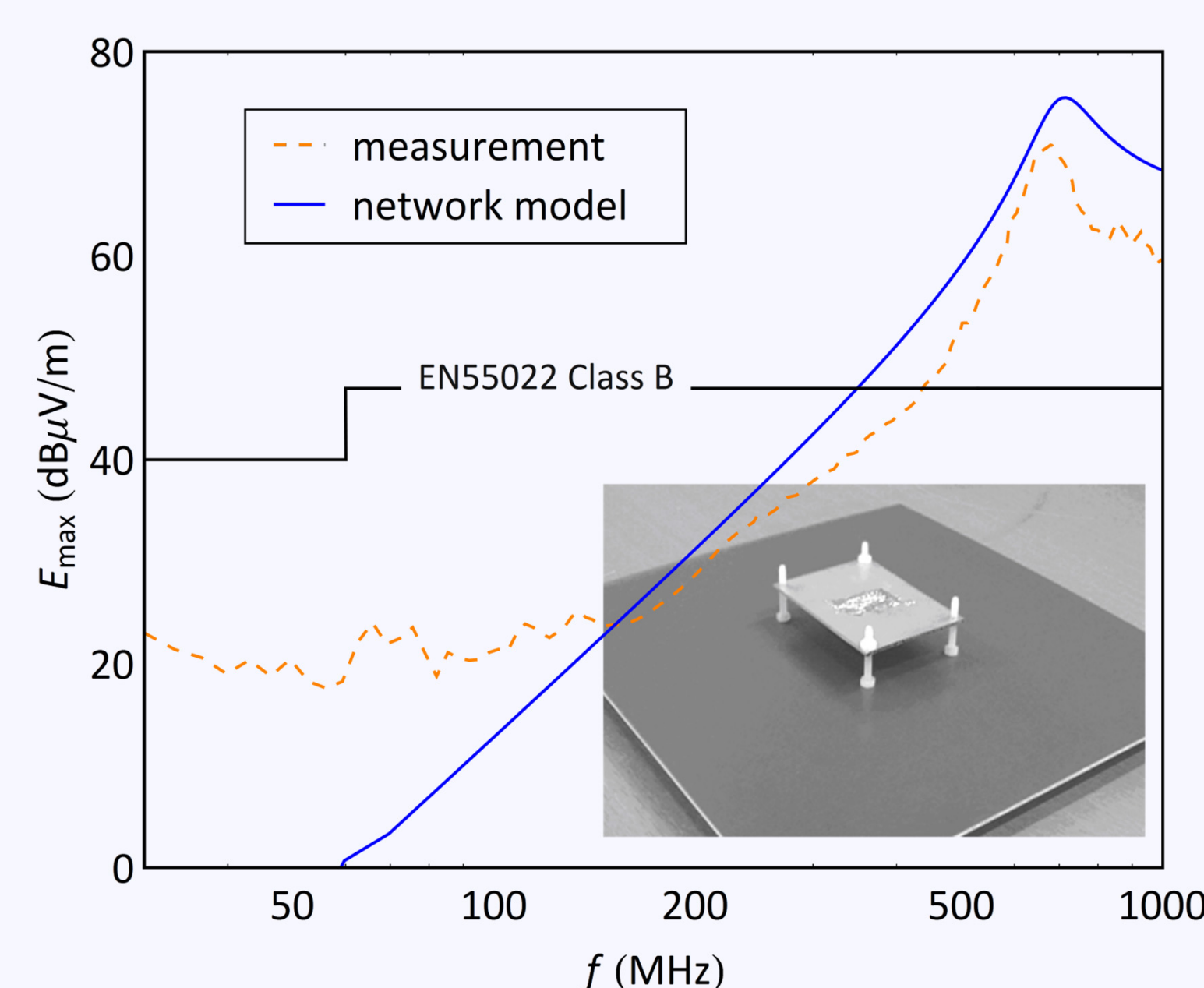
$$f_{res} \approx \frac{1}{2\pi \sqrt{L_{tot} C_{ant}}}$$

Parallel connection of all signal return inductances: $L_{tot} = \sum_{ij} L_{ij}^{-1}$

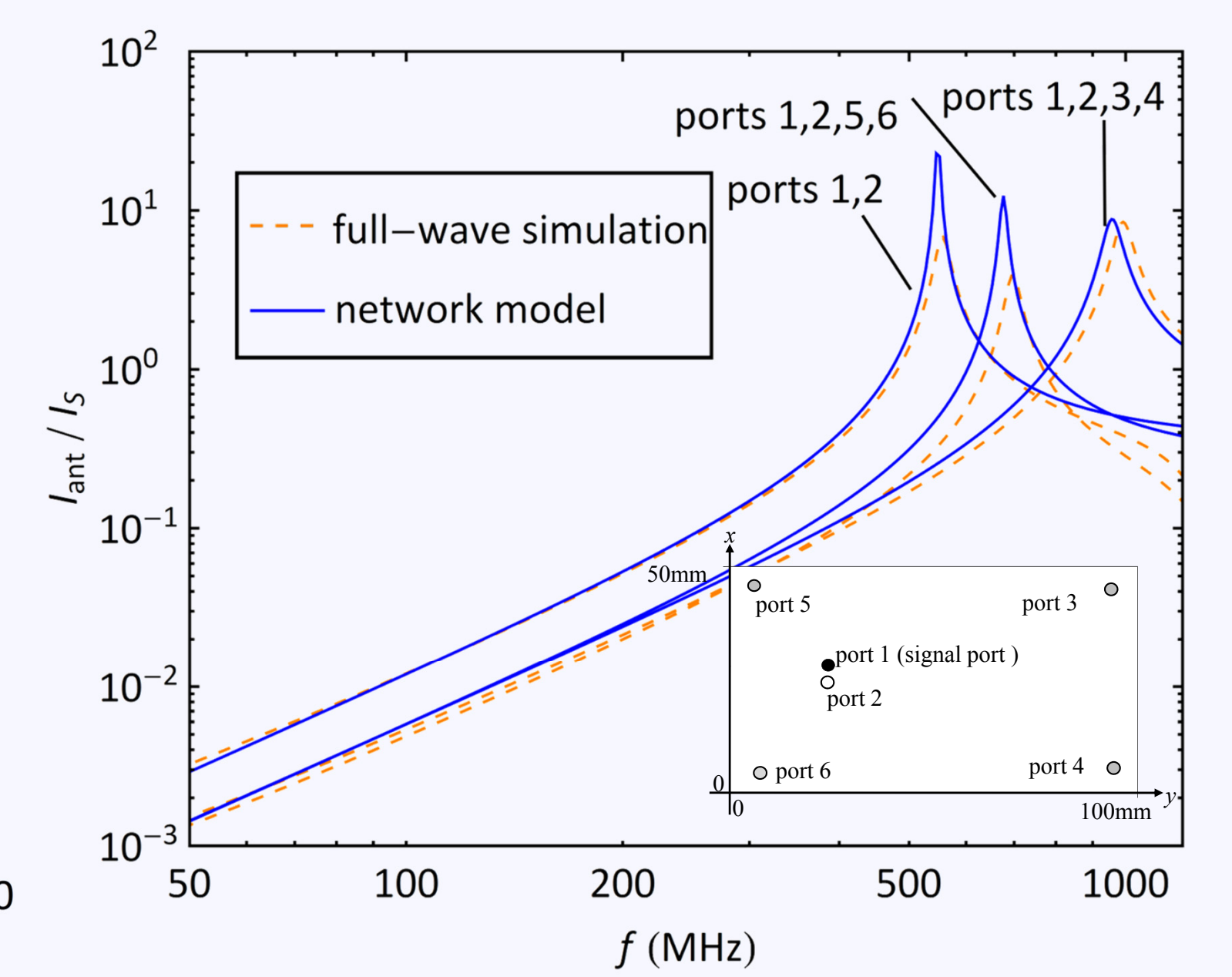
Reduced resonant network



RESULTS AND CONCLUSION



Frequency response of the maximum radiated field strength



Frequency response of the common-mode (antenna) current for different combinations

- Validation of network model by full-wave simulation and measurement
- Resonance frequency can be estimated in advance
- Accurate analytical expressions for equivalent-circuit elements provide physical insight and low computational effort